

# Drawbacks and Possible Improvements of Short Pulse Propagation Technique

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# Outline

- Introduction
- Drawbacks of the original SPP technique
- Possible modifications: SPP Light with TDT and S-parameters
- Sensitivity to test fixture - numerical experiment
- Examples of low-cost practical model identification
- Example of model identification up to 50 GHz
- Conclusion

# Introduction

- Design of PCB and packaging interconnects for data links running at bitrates 30 Gbps and beyond is challenging
  - Boards are not manufactured as designed
  - Making accurate measurements from DC to 50 GHz is very difficult
  - Accurate modeling over frequency bandwidth from DC to 50 GHz is difficult and even not possible in most of the EDA tools
- To have consistency in modeling and manufacturing, the same technique must be used at the material model identification and production model validation stages

Design success “fire triangle”



# Material models - specs

1GHz ; IPC TM650-2.5.5.9

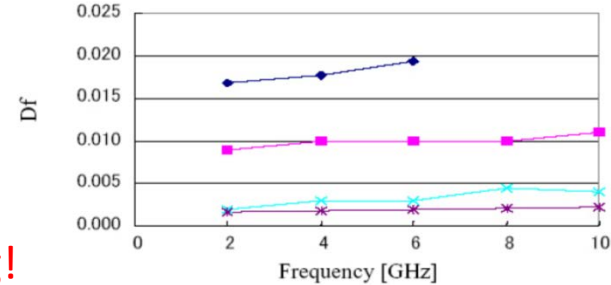
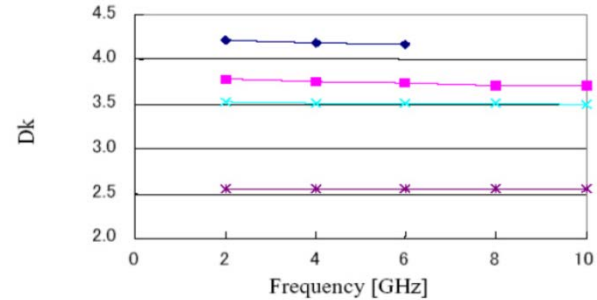
6-50GHz ; The method by H. Kawabata, Proceedings of the 36th European Microwave Conference, 388-391 (2006)

Core Type	Actual Thickness		Cloth Style	ply	Typical Resin Content (%)	Typical Dk									
	mil	mm				1GHz	6GHz	12GHz	18GHz	23GHz	29GHz	34GHz	40GHz	45GHz	50GHz
2	2.0	0.050	1035	1	67	3.25	3.23	3.22	3.21	3.21	3.21	3.21	3.21	3.21	3.21
2.6	2.6	0.065	1078	1	59	3.37	3.33	3.31	3.31	3.30	3.30	3.30	3.30	3.30	3.30
3	3.0	0.075	1078	1	65	3.28	3.25	3.24	3.23	3.23	3.23	3.23	3.23	3.23	3.23

Core Type	Actual Thickness		Cloth Style	ply	Typical Resin Content (%)	Typical Df									
	mil	mm				1GHz	6GHz	12GHz	18GHz	23GHz	29GHz	34GHz	40GHz	45GHz	50GHz
2	2.0	0.050	1035	1	67	0.002	0.003	0.004	0.004	0.004	0.004	0.005	0.005	0.005	0.005
2.6	2.6	0.065	1078	1	59	0.002	0.003	0.004	0.004	0.004	0.004	0.004	0.005	0.005	0.005
3	3.0	0.075	1078	1	65	0.002	0.003	0.004	0.004	0.004	0.004	0.005	0.005	0.005	0.005

\*The data above show actual values and are not guaranteed.

Not suitable directly for broadband modeling!  
Nothing for conductor roughness!!!



# Material models - needed

- **PCB dielectric models:**

Wideband Debye (aka Djordjevic-Sarkar):

$$\varepsilon(f) = \varepsilon_r(\infty) + \frac{\varepsilon_{rd}}{(m_2 - m_1) \cdot \ln(10)} \cdot \ln \left[ \frac{10^{m_2} + if}{10^{m_1} + if} \right]$$

Continuous-spectrum model  
Requires specification of DK and LT at one frequency point (2 parameters)

Multi-pole Debye: 
$$\varepsilon(f) = \varepsilon(\infty) + \sum_{n=1}^N \frac{\Delta\varepsilon_n}{1 + i \frac{f}{fr_n}}$$

Requires specification of value at infinity and poles/residues or DK and LT at multiple frequency points (more than 2 parameters)

- **PCB conductor surface roughness models:**

Modified Hammerstad (2 parameters):

$$K_{rh} = 1 + \left( \frac{2}{\pi} \cdot \arctan \left[ 1.4 \left( \frac{\Delta}{\delta} \right)^2 \right] \right) \cdot (RF - 1)$$

Huray snowball (1-ball, 2 parameters):

$$K_{rhu} = 1 + \left( \frac{N \cdot 4\pi \cdot r^2}{A_{hex}} \right) / \left( 1 + \frac{\delta}{r} + \frac{\delta^2}{2 \cdot r^2} \right)$$

- **Parameters for the models are not available and must be identified**

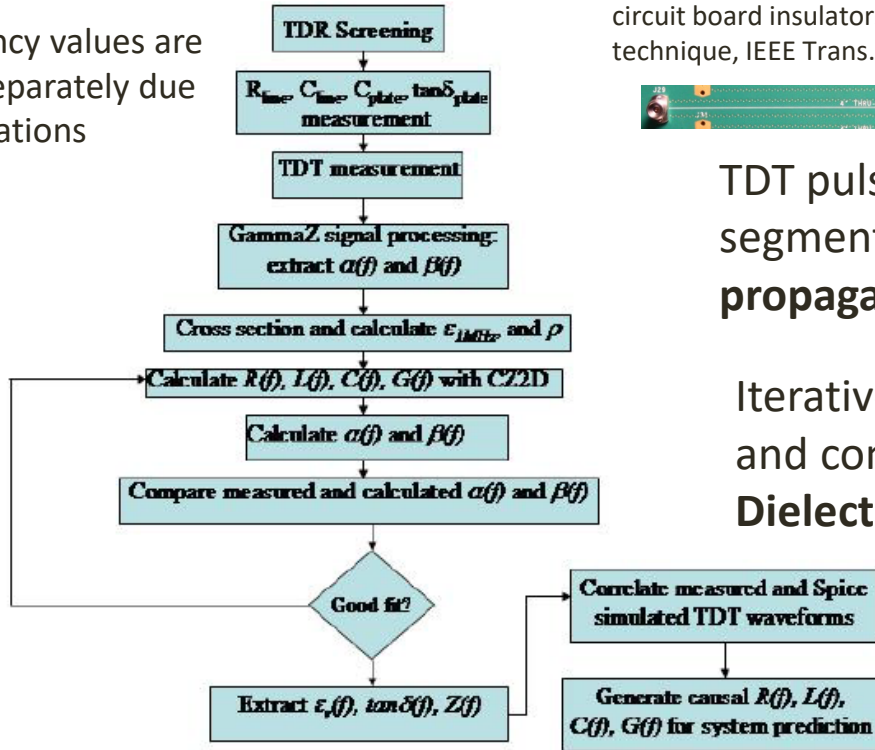
# Possible low-cost identification options

- Delta-L – uncertainty due to dependency on all reflections, uses S-parameters (requires VNA + measurement skills)
- Complete de-embedding (TRL, AFR, ISD,...) – unnecessary complicated – VNA, test fixture S-parameters are not needed,...
- Identification with GMS-parameters – similar to SPP with the improvements suggested here and at EPEPS'2015 (Shlepnev...)
- **Short Pulse Propagation – allows both model building and validation – standardized by IPC (IPC-TM-650 #2.5.5.12), simple, but may be confusing...**

# Step-by-Step Procedure for Short-Pulse-Propagation-Based Complex Permittivity Extraction

The following flowchart summarizes the extraction process:

Low frequency values are identified separately due to TDT limitations



A. Deutsch, T.-M. Winkel, G. V. Kopcsay, C. W. Surovic, B. J. Rubin, G. A. Katopis, B. J. Chamberlin, R. S. Krabbenhoft, Extraction of  $\epsilon_r$  and  $\tan \delta$  for printed circuit board insulators up to 30 GHz using the short-pulse propagation technique, IEEE Trans. on Adv. Packaging, vol. 28, 2005, N 1, p. 4-12.



TDT pulse responses of 2 line segments -> **Gamma (complex propagation constant)**

Iterative matching of measured and computed Gamma -> **Dielectric and Roughness Models**

# SPP Original: Steps 1-5

- **Step 1:** Select samples for the identification with TDR screening
  - Important step, sensitivity to the impedance variation should be investigated
- **Step 2:** Measure resistance of a t-line segment and capacitance of large disk at DC
  - Not needed - disk test can be eliminated by use of Wideband Debye model, resistance measurement can be replaced by matching attenuation at lower frequencies
- **Step 3:** Measure TDT step response, identify delay and construct preliminary transmission line model. – Not needed – eliminated by use of broadband models
- **Step 4:** Measure short pulse TDT for two line segments with about three to one ratio of the lengths.
  - TDT step should be allowed at this step to simplify the process, can be combined with step 1.
- **Step 5:** Convert short pulse TDT responses into complex propagation constant ( $\Gamma$ ) with windowing, Fourier transform and simple formula.
  - Can be improved by  $\Gamma$  extraction from S-parameters



# SPP Original: Steps 6-9

- **Step 6:** Cross-section test board and take measurements of the test structures
  - Most important step if the accuracy is a concern, sensitivity of Gamma to the incorrect values of the cross section should be investigated
- **Step 7:** With the measured dimensions, compute dielectric constant from the disk capacitance and resistivity of conductor from resistance, both measured at DC in step 2. – Not needed with the broadband Debye model.
- **Step 8:** Create transmission line model with multipole Debye dielectric model and tune parameters in dielectric model to match the measured attenuation and phase constant (real and imaginary parts of Gamma). –
  - Main material model extraction step - requires an accurate field solver and use of Wideband Debye dielectric model and Huray or modified Hammerstad roughness
- **Step 9:** Compute dielectric constant and loss tangent (includes roughness)
  - Needed only in case if the roughness model in the field solver is not reusable and has to be converted into the effective loss tangent for other EDA tools

# SPP Light with TDT

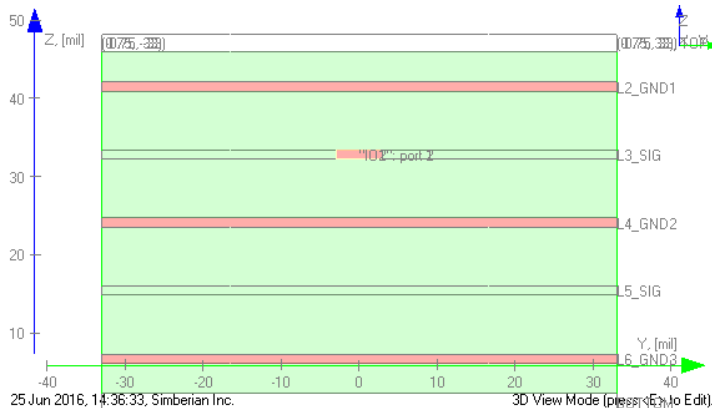
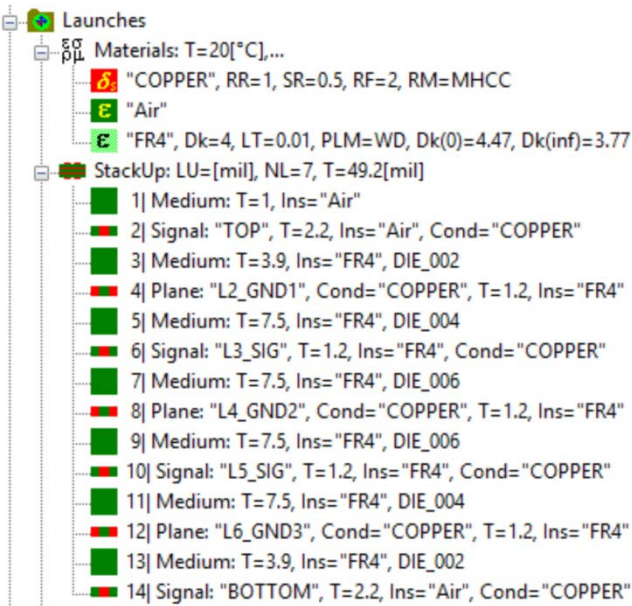
- **Step a:** Measure TDR and TDT step responses of line segments – select TDT responses of two segments with the close TDR impedances (strict guidance is needed)
- **Step b:** Convert TDT into short pulse response, window it and extract Gamma following the original technique
- **Step c:** Optionally, cross-section the board traces and measure the dimensions, to improve accuracy
  - Additional test fixtures can be considered as the alternative to the cross-sectioning
- **Step d:** Use field solver to build cross-section model matching Gamma extracted in step b) following either the original procedure or as in GMS technique (to separate conductor and dielectric losses)

# SPP Light with S-parameters

- **Step a:** Measure S-parameters of line segments, estimate quality and compute TDR and select two segments with the close TDR impedances (similar to GSM-parameters)
- **Step b:** Convert S-parameters into the reflection-less GSM-parameters and extract Gamma – see details at [EPEPS'2015]
- **Step c:** Optionally, cross-section the board traces and measure the dimensions, to improve accuracy
  - Additional test fixtures can be considered as the alternative to the cross-sectioning
- **Step d:** Use field solver to build cross-section model matching Gamma extracted in step b) following either the original procedure or as in GSM technique (to separate conductor and dielectric losses)

# Numerical experiment

6 mil strip line, 2 and 6 inch segments



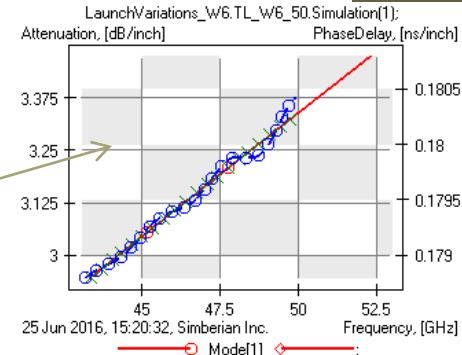
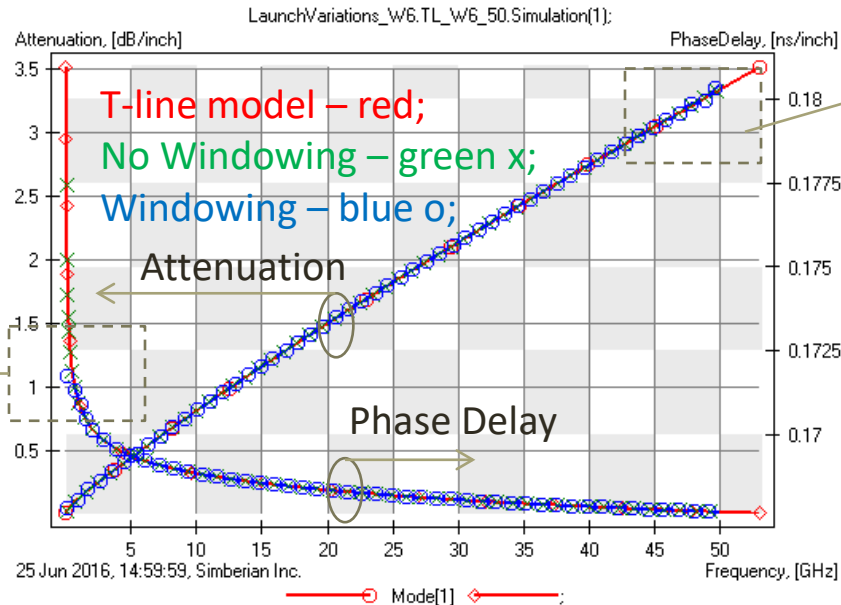
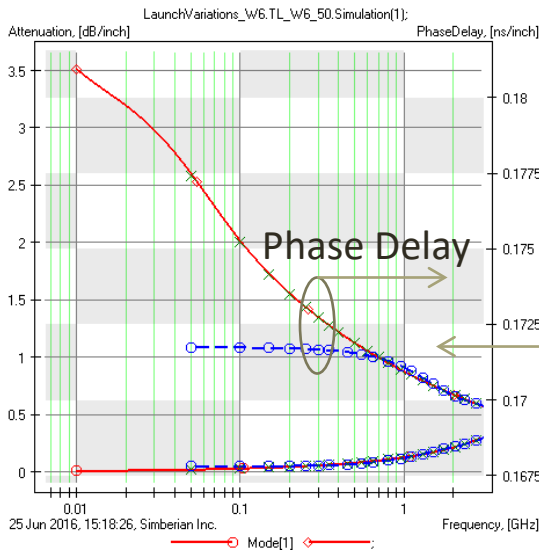
Dielectric: Wideband Debye, Dk=4, LT=0.01 @ 1 GHz;  
Conductor roughness: MHCC, SR=0.5, RF=1

Modelled with Simbeor THz

# Gamma extraction from TDT – 2 and 6 inch segments, no launches

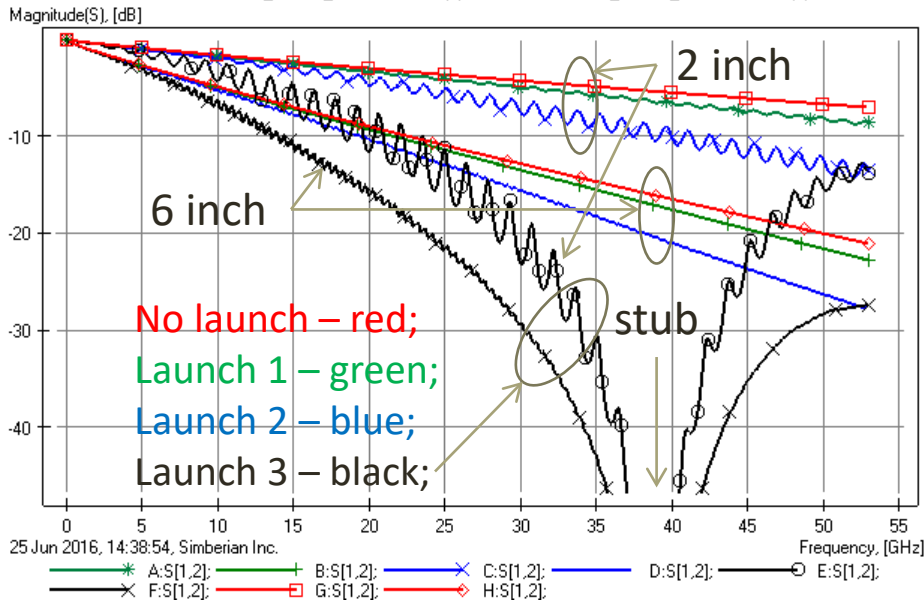
Windowing defect at low frequencies – wrong DC

Windowing defect at high frequencies - oscillations

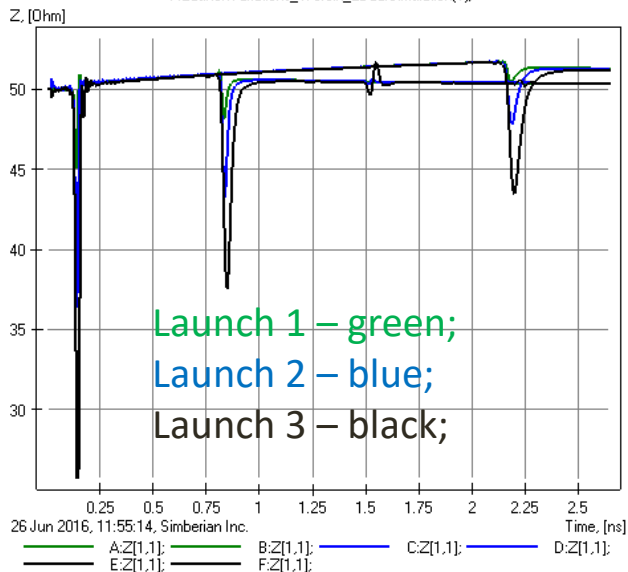


# 3 test pairs with different launches

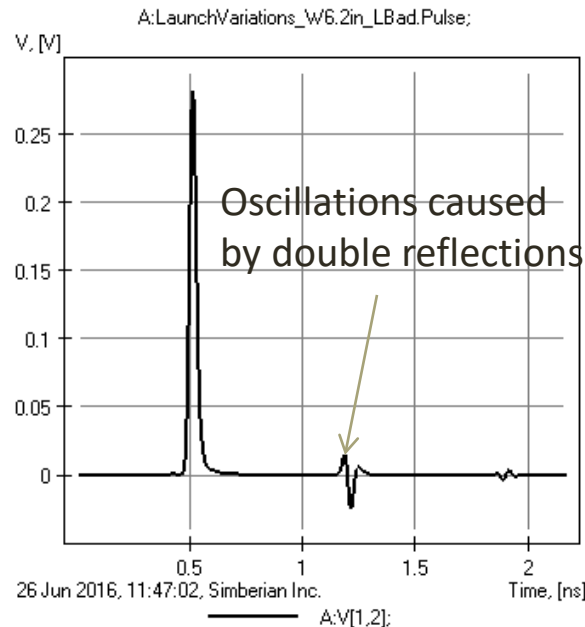
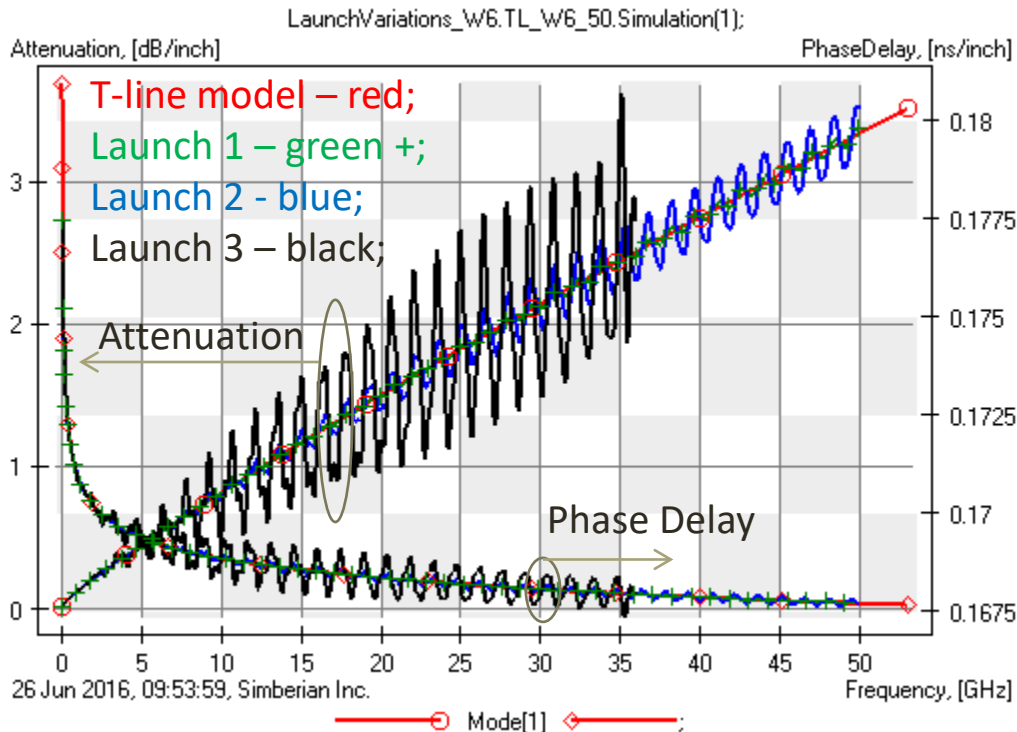
A: LaunchVariations\_W6.2in\_LGood.Simulation(1); B: LaunchVariations\_W6.6in\_LGood.Simulation(1);  
 C: LaunchVariations\_W6.2in\_LOk.Simulation(1); D: LaunchVariations\_W6.6in\_LOk.Simulation(1);  
 E: LaunchVariations\_W6.2in\_LBad.Simulation(1); F: LaunchVariations\_W6.6in\_LBad.Simulation(1);  
 G: LaunchVariations\_W6.2in\_Ideal.Simulation(1); H: LaunchVariations\_W6.6in\_Ideal.Simulation(1);



A: LaunchVariations\_W6.2in\_LGood.Simulation(1);  
 B: LaunchVariations\_W6.6in\_LGood.Simulation(1); C: LaunchVariations\_W6.2in\_LOk.Simulation(1);  
 D: LaunchVariations\_W6.6in\_LOk.Simulation(1); E: LaunchVariations\_W6.2in\_LBad.Simulation(1);  
 F: LaunchVariations\_W6.6in\_LBad.Simulation(1);

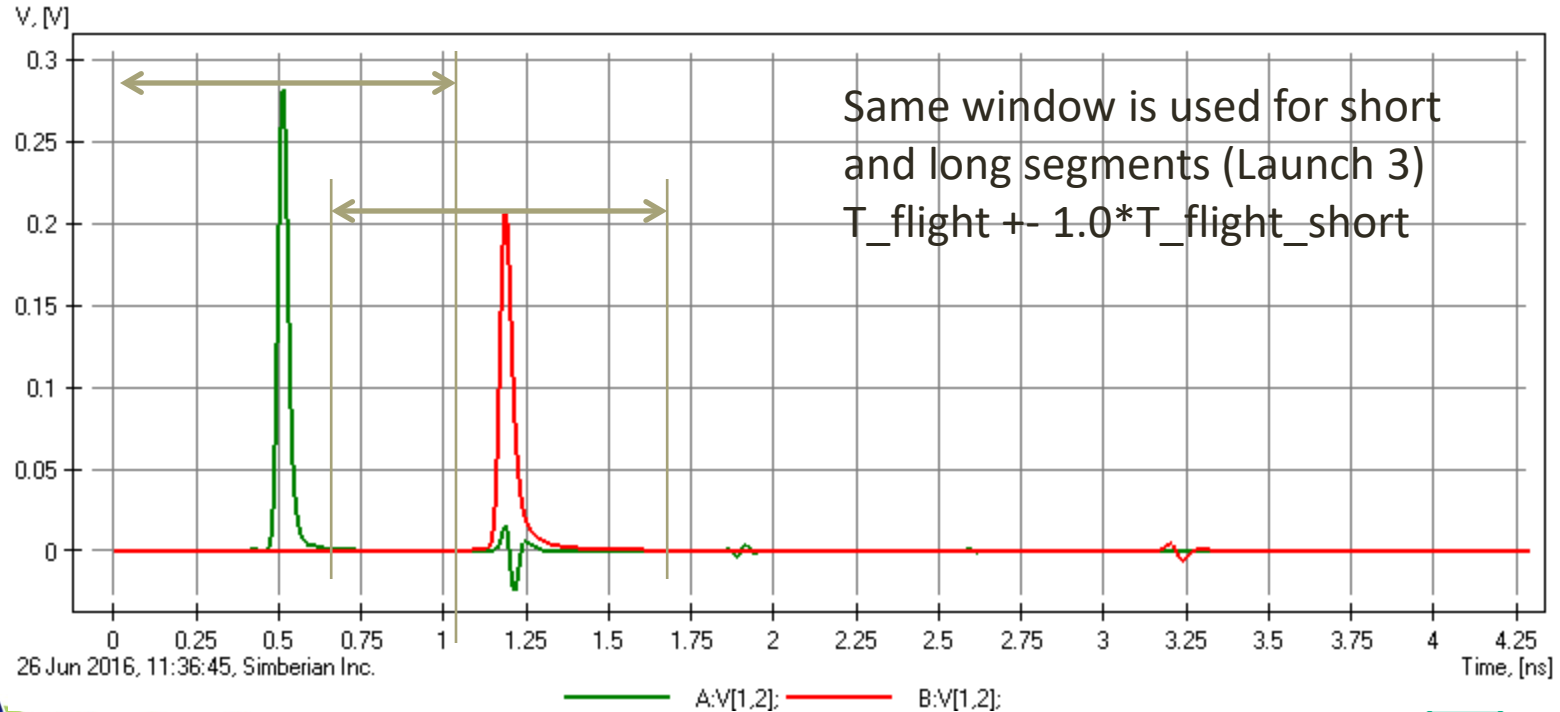


# Gamma extraction from TDT without windowing



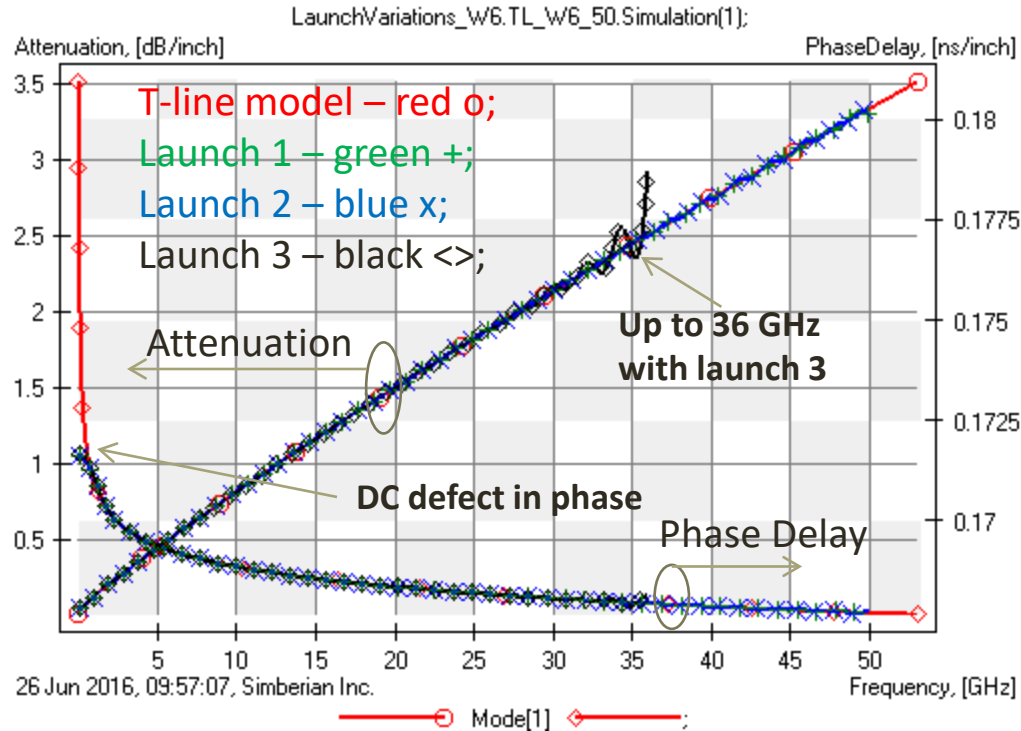
# Windowing to eliminate double reflections (de-embedding)

A:LaunchVariations\_W6.2in\_LBad.Pulse; B:LaunchVariations\_W6.6in\_LBad.Pulse;

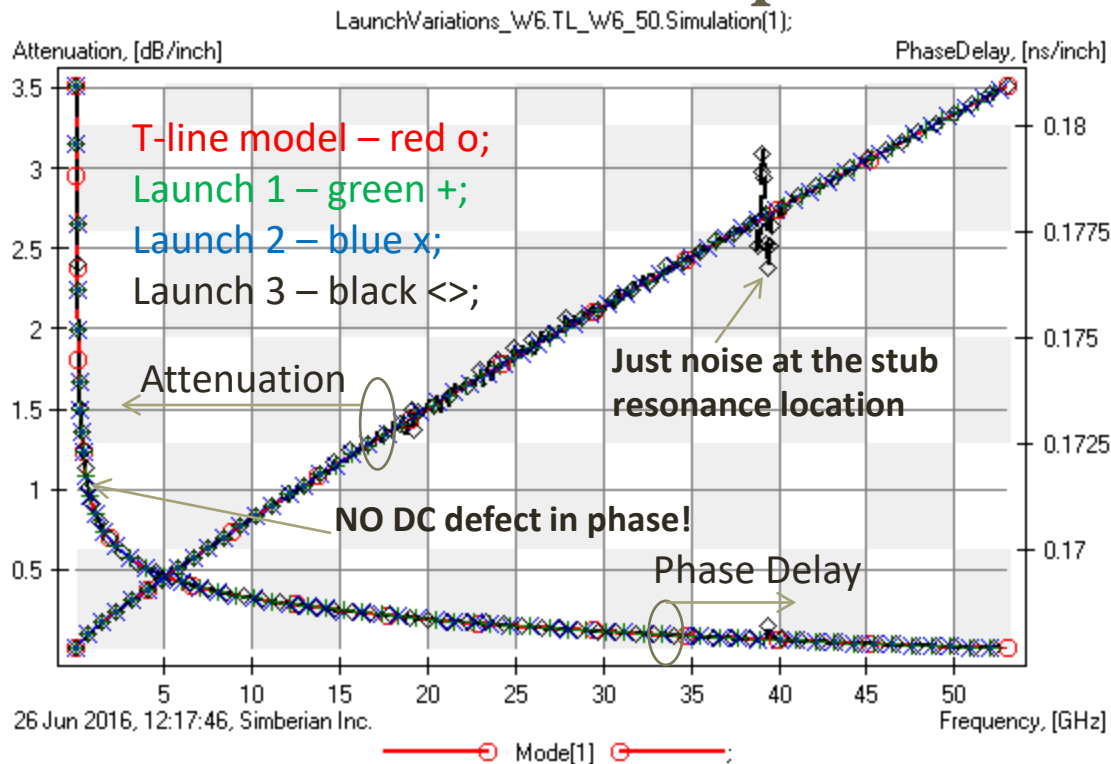




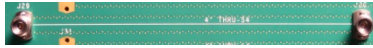
# Gamma extraction from TDT with windowing



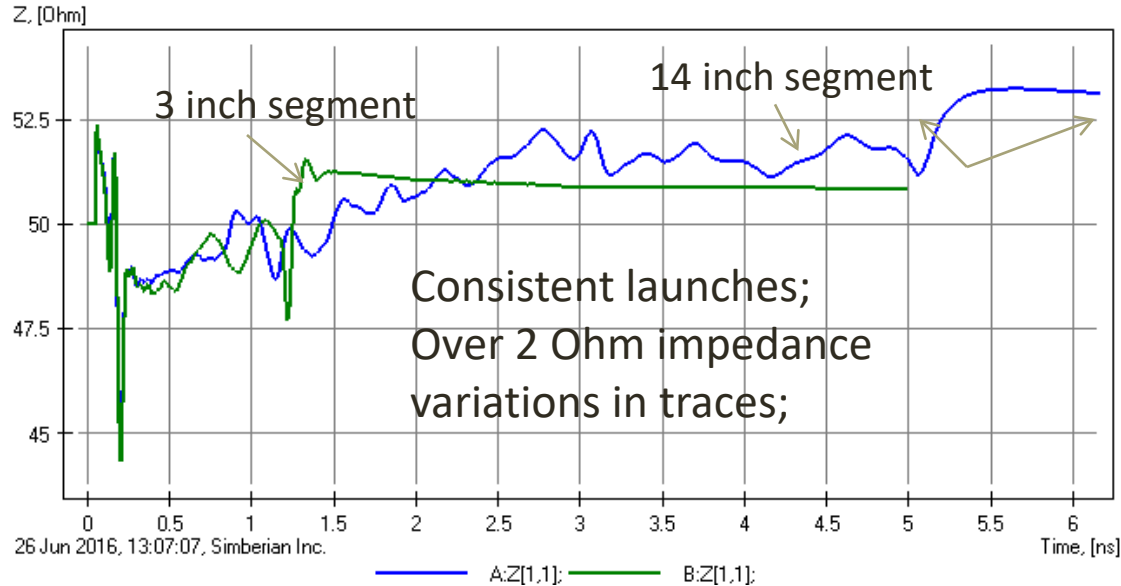
# Gamma extraction from S-parameters



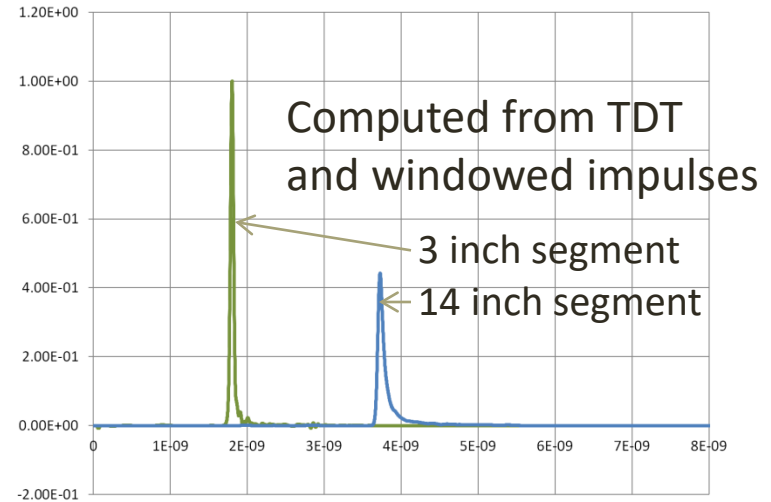
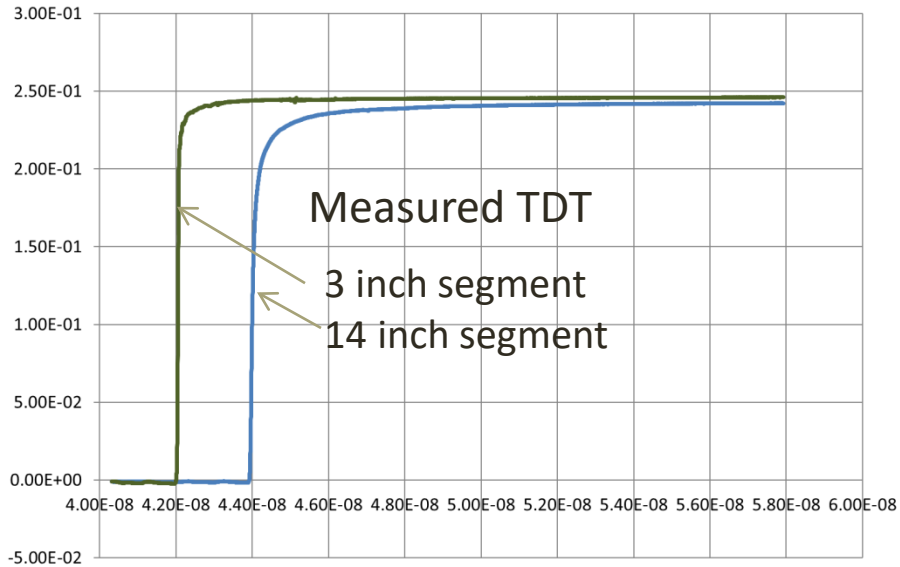
# Test board with SMA connectors - TDR



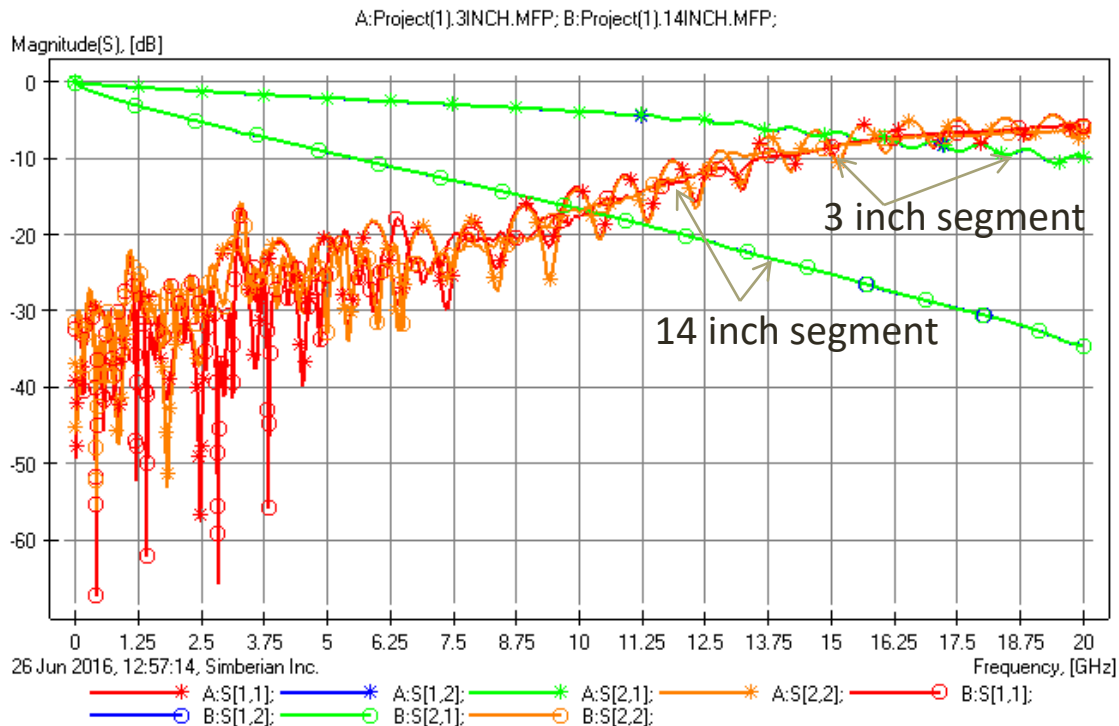
A:Project(1).14INCH.MFP; B:Project(1).3INCH.MFP;



# Test board with SMA connectors – TDT measurement

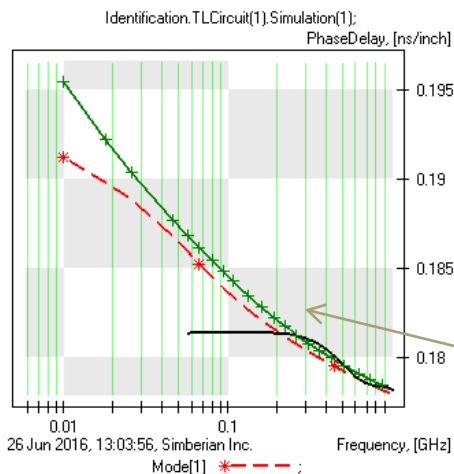


# Test board with SMA connectors – S-parameters measurements

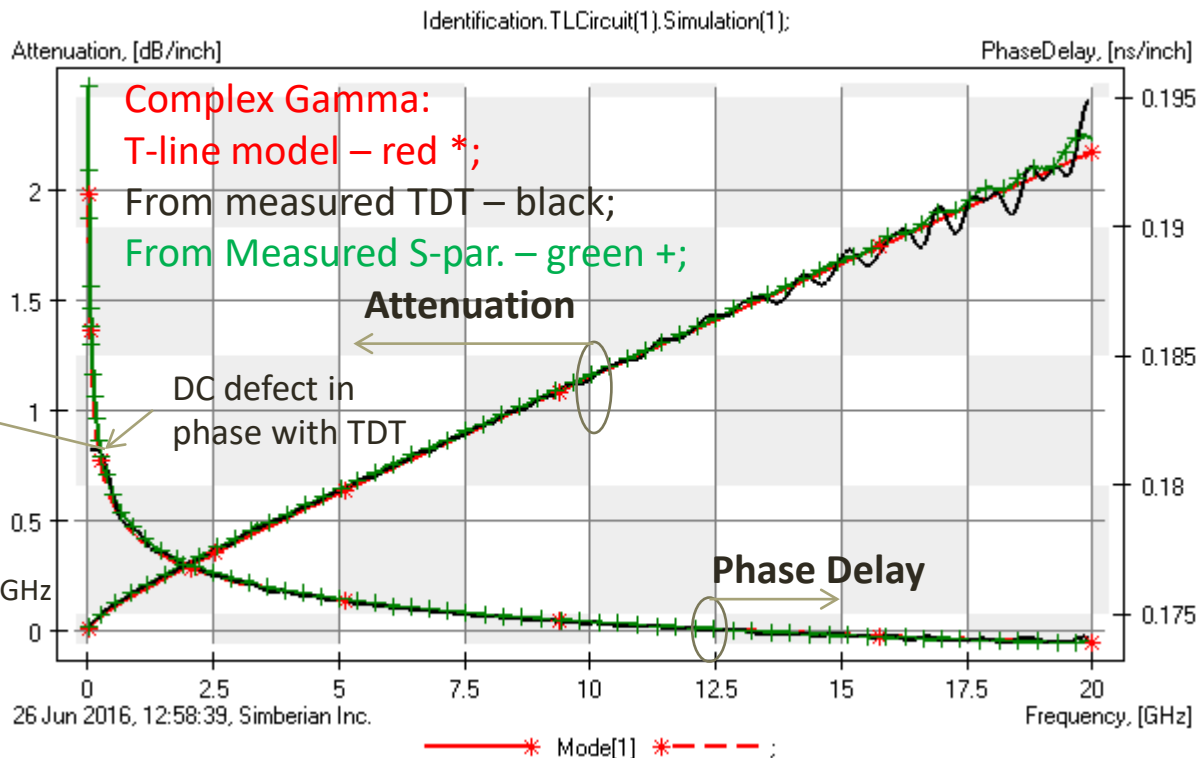


# Identification on test board with SMAs

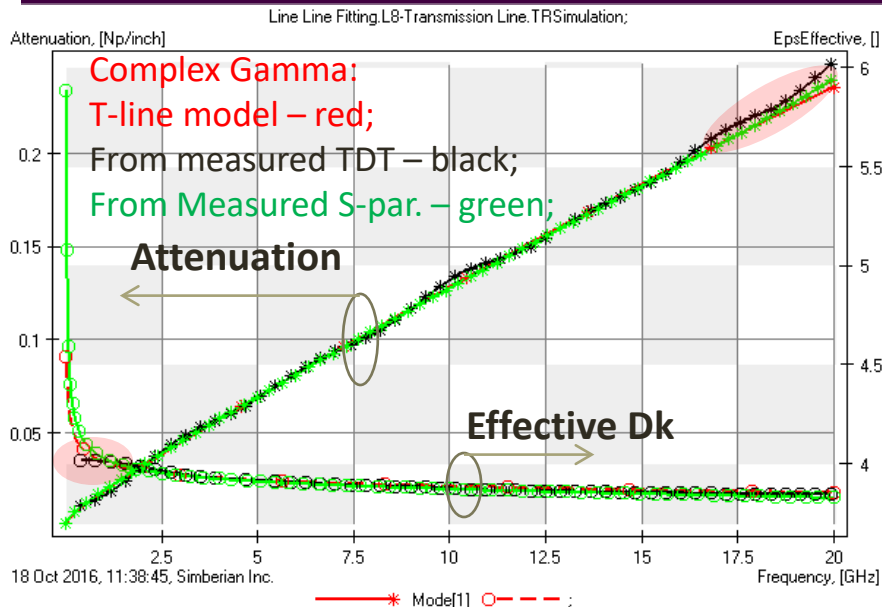
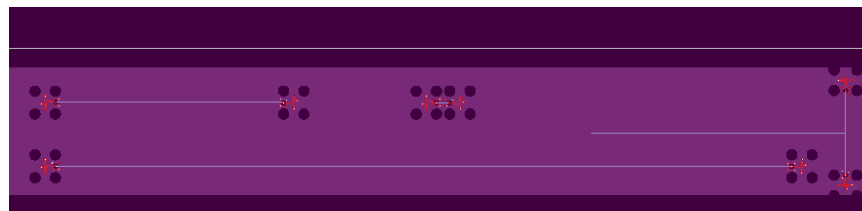
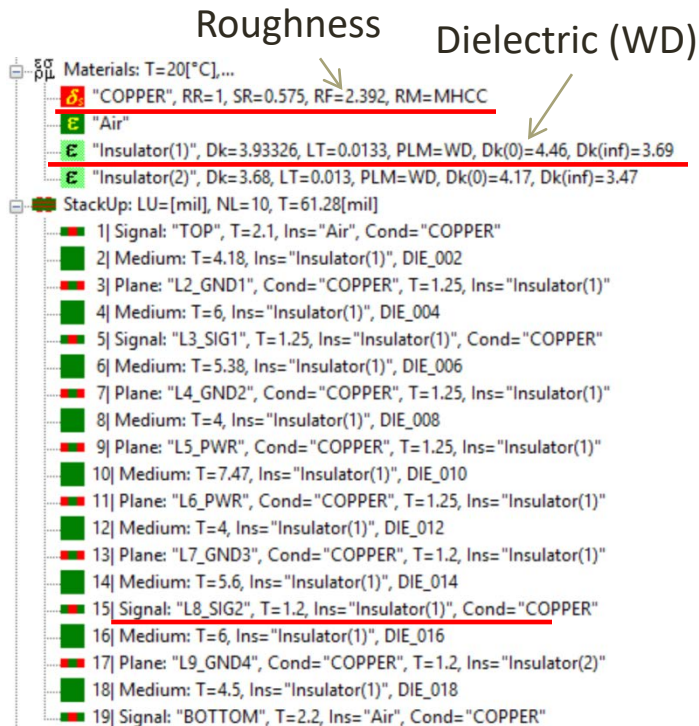
## DC defect



Dielectric: WD, Dk=4.33, LT=0.017 @ 0.5 GHz  
Roughness: MHCC, SR=0.3, RF=2



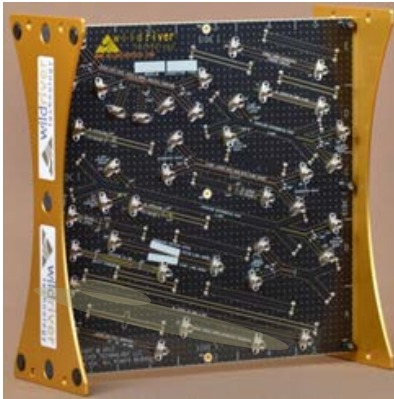
# Identification on test board with hand-held probes



Very inexpensive and quick!

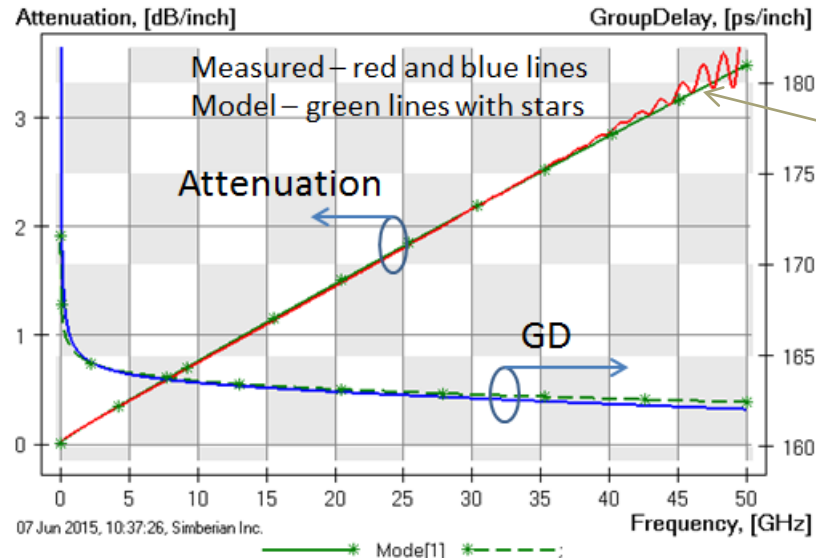
# Identification on CMP-28

- Dielectric: Wideband Debye dielectric model with  $Dk=3.8$  (3.66),  $LT=0.0117$  @ 1 GHz;
  - Conductor roughness: modified Hammerstad model with  $SR=0.32$   $\mu m$ ,  $RF=3.3$
- Models are usable even above 50 GHz!



CMP-28 channel modelling platform from Wild River Technology  
<http://www.wildrivertech.com/>

## Complex Gamma (SPP Light with S-parameters)





# Conclusion

- SPP technique has advantages over more complicated de-embedding techniques or simplistic like Delta-L – standardized, inexpensive, extendable, accurate, usable at design and manufacturing stages,...
  - SPP technique can be simplified from 9 to 4 steps (SPP Light)
  - SPP Light with S-parameters removes reflections and extends frequency range of SPP up to 50 GHz for PCB applications
- Further work – investigate sensitivity and improve dielectric and conductor roughness model separation with additional test structures